

DETERMINATION OF POLARIS AND OTHER SOURCES
AT LOW RADIO FREQUENCIES

BY

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A THESIS SUBMITTED TO THE GRADUATE COUNCIL
OF THE UNIVERSITY OF FLORIDA IN
PARTIAL FULFILLMENT OF THE REQUIREMENTS
FOR THE DEGREE OF MASTER OF SCIENCE

UNIVERSITY OF FLORIDA

1965

To Maria Angelina
Carolina
and Gabriela

ACKNOWLEDGMENTS

I wish to express my deepest gratitude to my research advisor, Dr. Thomas B. Carr, for his guidance and encouragement, and his patience in reading and correcting the manuscript. It has been a privilege to work under him in this project, and I have benefited in many ways as his student.

I am also grateful to Dr. Alex G. Smith for providing me with financial support and for serving as a committee member. I give my appreciation to Dr. John F. Oliver for the development of the necessary hardware and software, for his advice and for serving as a committee member.

My deep appreciation goes to Dean Claudio Anguila and Professor Hugo Moreno of Facultad de Ciencias Físicas y Matemáticas of University of Chile for their encouragement and interest in my career and for providing me with financial support. I am also grateful to Jorge Ray and Juan Aparicio for their encouragement and opinion and for providing me with the data at 45 MHz.

I wish to thank Jorge Ley for his help in the operation and calibration of the NMR NMR array, Chris St. Cyr for his encouragement and help in the computational part and Wesley Greenman and Richard Flagg for their help with the electronics.

My thanks goes to Hector Alvarez, Mauricio Múñoz and
Pascual Siles of Elipé Radio Observatory for their help
with the 45 MHz data...

I am grateful to the Astronomy Department and the
Physics Department and in particular to Drs. Heinrich
Rindholz, Robert E. Lowcock, F. Eugene Dennis, Charles F.
Repper and Richard H. Dettell for their financial support
in the form of teaching assistantships and fee waivers.

My gratitude goes to Mrs. Brenda Roberts for her
patience in correcting my English and for her devotion in
typing and editing the manuscript.

The realization of this project has been possible
thanks to the continuous support provided to the radio
astronomy program of the Department of Astronomy of the
University of Florida and to the Elipé Radio-Observatory
through a grant from National Science Foundation (Principal
Investigator Dr. Allen G. Reilly).

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Abstract of Thesis Presented to the Graduate School
of the University of Florida in Partial Fulfillment of the
Requirements for the Degree of Master of Science

OSCILLATION OF POLARONS AND OTHER SPECIES
AT LOW AUDIO FREQUENCIES

by

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December 1960

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Major Department: Astronomy

The investigation of the radio emission spectra of
pulsars at low frequencies is important since it helps to
clarify the mechanism responsible for this emission. At the
present, the low frequency end of the spectra is poorly
determined compared with more precise measurements avail-
able at higher frequencies. Also, measurements of pulsar
intensity variations at low frequencies due to scintilla-
tion are of value in the study of the interstellar medium.

Detection of pulsars at decimeter wavelengths is
difficult because of the deterioration of the signal-
to-noise ratio caused by the high galactic background temper-
tures and the tail-over of the spectra of most pulsars.
Also the dispersion of the pulses imposes the use of narrow
bandwidth, reducing the sensitivity of the system.

Two pulsars have been observed at low radio frequencies using narrow bandwidth to avoid pulse dispersion problems. Pulsar PSR1133+14 was observed at 24.3 MHz using the University of Florida Large array; pulsar PSR1545-14 was observed at 45 MHz using the Bigeye Radioobservatory Large array (CHLH). Long term integration techniques were used to reduce the data.

Both pulsars were successfully detected and the average pulse intensity and integrated pulse profile were obtained. Time resolutions of $P/30$ were obtained for PSR1133+14 and $P/15$ for PSR1545-14, where P is the pulse repetition period. The two observed pulsars were selected because of their low dispersion measure, long period (>1 sec), high pulse intensity and their location with respect to the galactic background.

Several other radio sources were also observed and were used to calibrate the effectiveness of the 24.3 MHz antenna.


THOMAS O. CARR

CHAPTER 1 INTRODUCTION

Pulsars were discovered in 1947 by J. Jocelyn Bell and Anthony Hewish. The first pulsar paper was published in Nature in February 1948 by Menzies et al. (1) telling about the discovery of pulsar PSR0318+41 and giving some of its parameters. Since then, about 141 pulsars have been detected and cataloged.

According to currently accepted models, pulsars are rapidly spinning neutron stars originating from supernova explosions. They are believed to have strong magnetic fields on the order of 10^{12} gauss, probably the strongest known magnetic fields in the Universe. The neutron star is believed to emit continuously a narrow beam of radio waves which coincides with the star, producing the observed pulses each time the beam sweeps past the earth. The periods of the pulsars which have been observed range from about 0.01 to 4.3 seconds. So far, for only 3 pulsars has it been possible to find the optical counterparts of a pulsar.

Several mechanisms have been proposed to explain the origin of the radio emission and how the rotational energy of the neutron star is converted into electromagnetic wave pulses; this subject still remains one of the least understood aspects of pulsars. Some of the important parameters

of a pulsar that could be measured. Radio emission models of the radio emission mechanism can be developed and the spectrum of the radiation, the pulse shape, peak intensity, and the time variations of these parameters. A large number of measurements have been made in the range of frequencies between 100 and 3,000 MHz, but only a few below 100 MHz (ii). The low frequency part of the spectrum is particularly interesting because most pulsars present a turnover of the spectrum at metric and decimetric wavelengths, and also because pulse shape and widths show some peculiarities in this region. The spectral maximum for most pulsars is in the range 80 to 250 MHz; this peak is nearly symmetrical with respect to the frequency at which the maximum occurs (ii). Measurements of the low frequency part of the pulsar spectra have been made by Bash et al. at 30 MHz (i), Brack and Gerasimko at 10, 10.5, 20 and 25 MHz (4, 5), Craft and Ormrod at 40 MHz (6) and more recently by Jurekian et al. at 41 and 107.5 MHz (7, 8).

Most pulsars observed at low frequencies are in that part of the sky which is observable from northern hemisphere observatories. Observations at the lowest frequencies have been carried out in the southern hemisphere only by Australian observatories, at 80 and 150 MHz (9, 10). Detection of pulsars at low radio frequencies is difficult because of the deterioration of the signal-to-noise ratio caused by the high galactic background temperatures, and the low frequency spectral turn-over exhibited by most

points. Dispersion in the atmospheric medium becomes more severe the lower the frequency. At higher frequencies, compensation can be made for dispersion, permitting the use of relatively high bandwidths. However, at the lower frequencies, it is generally necessary to restrict the bandwidth because of dispersion, further reducing the sensitivity of the detection system. Due to the intrinsic short pulse duration of most pulsars (100 msec), short time constants must be used, imposing still another limitation on the sensitivity of the system.

Nevertheless, the sensitivity of the system can be improved by averaging many independent records of the same source. If N records are averaged, the signal-to-noise ratio (S/N) is improved by a factor of \sqrt{N} . For pulsars, this can be accomplished by averaging groups of consecutive periods. Digital sampling is generally used to make averaging. Due to the relatively short pulsar periods, the signal needs to be sampled at a fast rate in order to have enough time resolution to meet the requirements of the sampling theorem. Since it is not feasible to manipulate the large amounts of data generated in the digitizing process by hand, digital computing techniques must be used, providing an efficient way to average, process, store and plot that data.

In the past few years, the development of small and inexpensive microprocessors has made possible improved methods of pulsar data handling and processing. If the

period of a pulse is known to be a good accuracy. It is possible to average many periods (it typically is in the range of a thousand periods), greatly improving the S/N ratio.

The University of Florida Radio Observatory, located in Alachua County about 18 miles west of Gainesville, and the Magill Radioastronomical Observatory of the University of Chile, located 20 miles southwest of Santiago, have long been engaged in a cooperative program of low frequency radio astronomy. Although much of the effort over the years has gone into the study of Jupiter's radio emission, other decimeter-wavelength sources are also investigated. Each of the two observatories has a very large antenna array, in addition to very smaller radio telescope antennas. The Florida array consists of 148 dipoles operating at a center frequency of 26.1 MHz, while the Chile array has 128 dipoles and operates at a center frequency of 45 MHz. Both arrays were believed to be suitable for observing pulsars, although previous attempts to use them for this purpose had not been successful.

The goal of the research on which this thesis is based was to develop a low frequency pulsar data acquisition and reduction system, and to use it to detect and measure the important parameters of at least one pulsar with each of the two large arrays (i.e. the one in Florida and the one in Chile). This goal was achieved.

CHAPTER 3 DATA ACQUISITION AND REDUCTION

Two pulsars, PSR1213+16 and PSR2229-14, were selected from the lists in Appendix A2 for observation because of their low dispersion measure, relative long period (0.1 second), high pulse intensity and favorable location with respect to the galactic background. Pulsar PSR1213+16 was observed at 26.1 MHz using a bandwidth of 8 kHz, and pulsar PSR2229-14 was observed at 40 MHz with a bandwidth of 16 kHz.

The detected IF signals of the pulsars were recorded on one channel of a magnetic tape, and a time code generator signal on the other channel. The tapes were played back looking at 1 kHz down, the pulsar signal sampled, and the resulting values stored in the MIMIC Model 470 computer. The time code generator signal provides the timing pulses for the sampling and also indicates Universal Time (UT). The sampling of the signal was accomplished by a 12 bit analog/digital converter (A/D) and a 100 MHz discriminator. Pulsar pulse averaging is carried out by a computer program which averages consecutive periods of the sampled pulsar signal, previously stored in the MIMIC computer.

2.1. THE APPLICATION OF RAIRIS 16B1111-16

The pulsar 16B1111-16 was observed for 26 nights between April 15, 1979, and June 9, 1979, using the University of Florida 16.3 MHz large array, located at the University of Florida Radio Observatory in Alachua County. The 16.3 MHz array is a nearly filled rectangular array of 648 half-wave dipoles that covers about 30,000 m². The half-power beamwidth (HPBW) is 1.5° north-to-south by 4.0° east-to-west. It is a phase steered array with full N-S directional control of the beam in quarter-beamwidth increments, and a selection of 8 E-W beam directions spaced one beamwidth apart. Phasing is accomplished by means of a network of Butler matrices, hybrid rings and plug-in phasing cables.

North-south phasing of the array for a source having a given declination is accomplished in two ways, a coarse adjustment in large increments of declination and a fine adjustment providing small steps. A change in the coarse phasing requires the replacement of many plug-in cables located throughout the array, and requires about 2 or 3 man-hours of labor. Fine phasing for the desired declination is quickly accomplished by means of switches at a single location. The beam pointing can be altered in small increments to any declination within a strip roughly 15° wide by means of the fine phasing controls. Pointing the beam to a new declination outside this strip requires a coarse phasing readjustment. The right ascension of the beam is of course

controlled by the rotation of the earth. Beams 19, 20, 21, 22, 23, 24, 25, 26, 27, 28, 29, 30, 31, 32, 33, 34, 35, 36, 37, 38, 39, 40, 41, 42, 43, 44, 45, 46, 47, 48, 49, 50, 51, 52, 53, 54, 55, 56, 57, 58, 59, 60, 61, 62, 63, 64, 65, 66, 67, 68, 69, 70, 71, 72, 73, 74, 75, 76, 77, 78, 79, 80, 81, 82, 83, 84, 85, 86, 87, 88, 89, 90, 91, 92, 93, 94, 95, 96, 97, 98, 99, 100, 101, 102, 103, 104, 105, 106, 107, 108, 109, 110, 111, 112, 113, 114, 115, 116, 117, 118, 119, 120, 121, 122, 123, 124, 125, 126, 127, 128, 129, 130, 131, 132, 133, 134, 135, 136, 137, 138, 139, 140, 141, 142, 143, 144, 145, 146, 147, 148, 149, 150, 151, 152, 153, 154, 155, 156, 157, 158, 159, 160, 161, 162, 163, 164, 165, 166, 167, 168, 169, 170, 171, 172, 173, 174, 175, 176, 177, 178, 179, 180, 181, 182, 183, 184, 185, 186, 187, 188, 189, 190, 191, 192, 193, 194, 195, 196, 197, 198, 199, 200, 201, 202, 203, 204, 205, 206, 207, 208, 209, 210, 211, 212, 213, 214, 215, 216, 217, 218, 219, 220, 221, 222, 223, 224, 225, 226, 227, 228, 229, 230, 231, 232, 233, 234, 235, 236, 237, 238, 239, 240, 241, 242, 243, 244, 245, 246, 247, 248, 249, 250, 251, 252, 253, 254, 255, 256, 257, 258, 259, 260, 261, 262, 263, 264, 265, 266, 267, 268, 269, 270, 271, 272, 273, 274, 275, 276, 277, 278, 279, 280, 281, 282, 283, 284, 285, 286, 287, 288, 289, 290, 291, 292, 293, 294, 295, 296, 297, 298, 299, 300, 301, 302, 303, 304, 305, 306, 307, 308, 309, 310, 311, 312, 313, 314, 315, 316, 317, 318, 319, 320, 321, 322, 323, 324, 325, 326, 327, 328, 329, 330, 331, 332, 333, 334, 335, 336, 337, 338, 339, 340, 341, 342, 343, 344, 345, 346, 347, 348, 349, 350, 351, 352, 353, 354, 355, 356, 357, 358, 359, 360, 361, 362, 363, 364, 365, 366, 367, 368, 369, 370, 371, 372, 373, 374, 375, 376, 377, 378, 379, 380, 381, 382, 383, 384, 385, 386, 387, 388, 389, 390, 391, 392, 393, 394, 395, 396, 397, 398, 399, 400, 401, 402, 403, 404, 405, 406, 407, 408, 409, 410, 411, 412, 413, 414, 415, 416, 417, 418, 419, 420, 421, 422, 423, 424, 425, 426, 427, 428, 429, 430, 431, 432, 433, 434, 435, 436, 437, 438, 439, 440, 441, 442, 443, 444, 445, 446, 447, 448, 449, 450, 451, 452, 453, 454, 455, 456, 457, 458, 459, 460, 461, 462, 463, 464, 465, 466, 467, 468, 469, 470, 471, 472, 473, 474, 475, 476, 477, 478, 479, 480, 481, 482, 483, 484, 485, 486, 487, 488, 489, 490, 491, 492, 493, 494, 495, 496, 497, 498, 499, 500, 501, 502, 503, 504, 505, 506, 507, 508, 509, 510, 511, 512, 513, 514, 515, 516, 517, 518, 519, 520, 521, 522, 523, 524, 525, 526, 527, 528, 529, 530, 531, 532, 533, 534, 535, 536, 537, 538, 539, 540, 541, 542, 543, 544, 545, 546, 547, 548, 549, 550, 551, 552, 553, 554, 555, 556, 557, 558, 559, 560, 561, 562, 563, 564, 565, 566, 567, 568, 569, 570, 571, 572, 573, 574, 575, 576, 577, 578, 579, 580, 581, 582, 583, 584, 585, 586, 587, 588, 589, 590, 591, 592, 593, 594, 595, 596, 597, 598, 599, 600, 601, 602, 603, 604, 605, 606, 607, 608, 609, 610, 611, 612, 613, 614, 615, 616, 617, 618, 619, 620, 621, 622, 623, 624, 625, 626, 627, 628, 629, 630, 631, 632, 633, 634, 635, 636, 637, 638, 639, 640, 641, 642, 643, 644, 645, 646, 647, 648, 649, 650, 651, 652, 653, 654, 655, 656, 657, 658, 659, 660, 661, 662, 663, 664, 665, 666, 667, 668, 669, 670, 671, 672, 673, 674, 675, 676, 677, 678, 679, 680, 681, 682, 683, 684, 685, 686, 687, 688, 689, 690, 691, 692, 693, 694, 695, 696, 697, 698, 699, 700, 701, 702, 703, 704, 705, 706, 707, 708, 709, 710, 711, 712, 713, 714, 715, 716, 717, 718, 719, 720, 721, 722, 723, 724, 725, 726, 727, 728, 729, 730, 731, 732, 733, 734, 735, 736, 737, 738, 739, 740, 741, 742, 743, 744, 745, 746, 747, 748, 749, 750, 751, 752, 753, 754, 755, 756, 757, 758, 759, 760, 761, 762, 763, 764, 765, 766, 767, 768, 769, 770, 771, 772, 773, 774, 775, 776, 777, 778, 779, 780, 781, 782, 783, 784, 785, 786, 787, 788, 789, 790, 791, 792, 793, 794, 795, 796, 797, 798, 799, 800, 801, 802, 803, 804, 805, 806, 807, 808, 809, 810, 811, 812, 813, 814, 815, 816, 817, 818, 819, 820, 821, 822, 823, 824, 825, 826, 827, 828, 829, 830, 831, 832, 833, 834, 835, 836, 837, 838, 839, 840, 841, 842, 843, 844, 845, 846, 847, 848, 849, 850, 851, 852, 853, 854, 855, 856, 857, 858, 859, 860, 861, 862, 863, 864, 865, 866, 867, 868, 869, 870, 871, 872, 873, 874, 875, 876, 877, 878, 879, 880, 881, 882, 883, 884, 885, 886, 887, 888, 889, 890, 891, 892, 893, 894, 895, 896, 897, 898, 899, 900, 901, 902, 903, 904, 905, 906, 907, 908, 909, 910, 911, 912, 913, 914, 915, 916, 917, 918, 919, 920, 921, 922, 923, 924, 925, 926, 927, 928, 929, 930, 931, 932, 933, 934, 935, 936, 937, 938, 939, 940, 941, 942, 943, 944, 945, 946, 947, 948, 949, 950, 951, 952, 953, 954, 955, 956, 957, 958, 959, 960, 961, 962, 963, 964, 965, 966, 967, 968, 969, 970, 971, 972, 973, 974, 975, 976, 977, 978, 979, 980, 981, 982, 983, 984, 985, 986, 987, 988, 989, 990, 991, 992, 993, 994, 995, 996, 997, 998, 999, 1000.

season at which time it was necessary to turn the clock down for the summer.

A block diagram for the interconnections of the equipment used to observe the pulsar is shown in Figure 2.1. The equipment basically included a Collins R 95B receiver, a Raytheonco 1001 magnetic tape recorder, a Tracer 50 frequency standard, and a Spence-Cowen E151 time code generator/reader.

The time code generator (TCG) was synchronized against WWV--this was accomplished by triggering an oscilloscope with the 1 pulse per second (1 PPS) output of the TCG and measuring the delay of WWV 1 PPS. The 1 PPS of the TCG was brought as close as possible to the 1 PPS of WWV. Usually between 5 to 20 ms by making short interruptions of the 1 kHz signal from the Tracer frequency standard. The delay between the two 1-PPS pulse trains was measured to an accuracy of about 10.0 ns, and from that information the D.T. was obtained, except for the propagation delay between Boulder, Colorado, and State County Radio Observatory. No corrections have been made for this delay because no attempt was made to print the epoch of this observation with any other observations.

The measured drift of the Tracer time standard is about 40 μ s/hr; this means a drift in the epoch of about 100 ns between pulsar transmits by hours 20 and 26 ($\sim 10^6$ sec²), this drift is small enough to be noticed, considering the time



Figure 3-1. Block diagram of the equipment used to record police File 113.104.

transmitted at 1000 Hz. P - (pulses) individual jth this pulse.

The detected IF output of the receiver was filtered on channel 1 of the Agilent at 15 MHz and the recording level for the pulsar background was set to about 1 V on the recording level monitor, to prevent any saturation during signal peaks.

The time code generator was recorded on channel 2 of the Agilent. This signal serves a double purpose: it provides a time reference for the pulsar signal, and later on, in the data reduction process, provides the timing signal for the EIE 1 microprocessor.

As a back-up to obtain the correct time, at the beginning of each observing session the WWV signals were recorded on channel 1 (for FOS) also on channel 2 for about 2-3 minutes. The correct time can be obtained by measuring the delay between the 1 PPS of WWV and the 1 PPS of the FOS.

The receiver was tuned to a center frequency of 84.5 MHz and the bandwidth set to 8 kHz. Using data for higher frequencies an extrapolated pulse width for the band of about 80 MHz was obtained. Pulse broadening due to dispersion $\Delta t = 8 \text{ pulses ns}^{-2}$ is about 18 ns (for a bandwidth $\Delta\nu = 8 \text{ kHz}$), which was considered acceptable.

2-2. Data Acquisition of Pulsar PSR1514-14

Pulsar PSR1514-14 was observed for 11 nights between May 18 and May 30, 1991, using the Westf Radio Astronomical

Superstation, 4.2 km (large array, located about 20 m apart) southwest of Santiago, Chile. The 45 MHz array is a circular hexapolar array of 128 half-wave dipoles with an effective area of $12,000 \text{ m}^2$. It is a transmit instrument with 8 independently phased subelements in the E-E direction and can be oriented to a zenith angle of about 45° in the meridian plane. The array is 2.1° north-to-south by 2.6° east-to-west. A helixpin beam system is generated by a Butler matrix. A more detailed description of the array has been given by Briggs (14) and Briggs et al. (15).

Pulsar PSR0437-28 was recorded for 16 equates each night, which is the time it takes to drift between the E-W 3 dB points of the beam. This pulsar was selected because of its relatively high pulse intensity ($14.2 \times 10^{-20} \text{ W m}^{-2} \text{ Hz}^{-1}$ at 14 MHz), long period (1.56 ms) and its low dispersion measure ($DM = 12 \text{ pulses cm}^{-3}$). A zenith angle of only 18° at transit was also another favorable choice considered.

The equipment used for the data acquisition of this pulsar consisted of a modified General Dynamics telemetry receiver, a Magneport 1800 magnetic tape recorder, a Sytron Sinter 8225 600 and a Rhode and Schwarz quartz time standard. The antennae elements are similar to those shown in Figure 2-1. During the observing sessions the pulsar was transmitting at dawn, and no interference was noticed. Five of the eleven nights gave reliable data, the other two nights were discarded because of equipment malfunctions.

A bandwidth of 18 kHz was used, this is the maximum bandwidth of the receiver. Using data obtained at higher frequencies, an extrapolated pulse pulse width at the base of about 110 ns was obtained at 85 MHz. The pulse broadening due to the 18 kHz bandwidth is 11 ns. This was considered acceptable.

The quartz time standard is checked daily against the Chilean National Observatory cesium time standard. The epoch is believed to be known with an accuracy of about 100 ns. The drift in time is about 1.4 $\mu\text{s}/\text{s}$ for about 8,40180 $\mu\text{s}/P$, where $P = 1.56 \mu\text{s}$, the pulse period. The drift is 1.2 ns is the 50 minutes between the next-lowest JPLR. For all practical purposes considered in this research, this is negligible.

3.2. Data Reduction

Data reduction for both pulvers was performed using the facilities of the Department of Astronomy of the University of Florida in Gainesville. The basic configuration and interconnections of the equipment for data reduction are shown in Figure 3-2. The schematics for the main relevant blocks are presented in Appendix C.

As stated earlier, the IF signal at the receiver is band limited to 8 or 18 kHz and appropriately smoothed. The resulting wide frequency signal is recorded in analog form on channel 1 of the Regulated tape recorder at a tape speed of 15 IPS. The time code generator is recorded on channel 2.

In order to achieve the desired time resolution, it will be able to sample the data at an appropriate rate, the magnetic tapes were played back at $3 \frac{1}{4}$ IPS, providing a slow down of 4:1.

Pulsar signals from channel 1 of the Magnetron are sent through a diode detector and smoothing filter to the 12 line A/D converter. The filter time constant was adjusted to 40 ns for both the rising and falling edges of the wave pulses.

The TOS signal from channel 2 of the Magnetron is connected to the input of the TOS working now as a counter to provide both a visual display of the time and a 1.775 pulse (taken from a modified output). The latter gives the synchronizing signal to start the sampling of the data at a precisely known time. It is particularly important to know the time at which the sampling is started in order to be able to compare in phase 1 or more blocks of data.

In order to obtain the timing signal for the KIM 1, the TOS signal from channel 2 of the Magnetron is connected also to a phase locked loop (PLL) circuit. The PLL circuit locks at a particular frequency of the signal, following only the peak of the frequencies present in the complex wave form of the TOS signal. The 1800 Hz signal of the TOS is transformed to 750 Hz due to the 4:1 slow down. The PLL was carefully tuned to 750 Hz, and the output then divided by 3. The resultant 183 Hz constitutes the timing signal for the KIM 1.

A gated DC output from the PLL circuit was monitored on a Brook Bank 11 chart recorder. This provides a good means

to detect any signal synchronization or dropout of the TOS playback signal. If there is any loss of synchronization between the input and TOS signals, then the gated DC output falls to zero, giving a visual indication to stop the sampling process. If synchronization were lost, the pulse signals would be averaged out of phase, wiping out any pulse at the end of the integration process. A means was provided for starting the sampling at the next 1 PPS pulse from the TOS occurring after a reset was thrown. This made it possible to have the exact time at which sampling started. This was accomplished by connecting the start switch to the D input and the 1-PPS signal to the clock (C) input of a D-type Flip-Flop. The output of the flip flop gates the 125 Hz signal in a NAND gate (see schematic in Appendix C). When the start switch is turned on, the next 1 PPS of the TOS enables the 125 Hz signal to go into the XMR 1 and the sampling is started.

The sampled values loaded in hexadecimal are sent through the Maxframe terminal to the Intel® 470 computer and stored there for further processing and analysis.

Since the maximum number of cards that can be stored under a new data file name is 400, it was necessary to break up the data into several "blocks" to meet this requirement. The 8 minutes of data obtained in one beam for pulsed FAIR1231+14 were broken up in 3 blocks of 4 minutes each, obtaining about 312 cards for each block (for a TRAP = 25 and timing signal frequency = 125 Hz, see Appendix B). The

18 minutes of data obtained each night for pulsar PSR0415-16 was broken up into 4 blocks of 5 minutes each, generating about 190 cards for each block.

The method of breaking up the data into blocks proved to be useful also because you can inspect the results of the average of each block separately, making it possible to search by eye for a possible pulse that might appear at the same location in the different blocks. It is highly unlikely that interference transients would appear in the same location in each of several blocks.

Monitoring of the DC gated output of the PLL circuit is very useful since synchronization is sometimes lost when a tape is played back. If that happens, then that portion of the tape is skipped and a new block started.

The K18-1 program which is necessary to digitize the data is loaded into the microprocessor from a cassette tape recorder.

2.4 Block Designation for Each Pulsar

Many blocks of data were generated in the data reduction process. In order to avoid any confusion with the data and to be able to distinguish quickly between blocks, a designation scheme was developed for referring to a specific data block. The designation format contains an abbreviated form of the pertinent information for that block. For pulsar PSR1510-18, the following block designation was adopted:



Additional beam designations are as follows:

- 00 2nd beam east of stream
- 01 1st beam east of stream
- 10 1st beam west of stream
- 02 2nd beam west of stream

In order to distinguish between the different blocks within a given beam, a capital letter was used, usually A and B. In case it became necessary to break up the data into more than two blocks (i.e. due to loss of synchronism of the PCC signal) C was also used.

The indicator for the date of the night on which the pulse was observed was an integer, as follows:

DATE	CODE
1	April 25, 1979
2	April 27, 1979
3	April 28, 1979
4	April 30, 1979
5	May 10, 1979
6	May 25, 1979
7	May 26, 1979
8	May 30, 1979

The designation scheme for the nights of data upon Rover PMS045-14 is slightly different, as follows:



The block designation is usually one of the letters A, B, C, D, E or F.

The nights were numbered as follows.

<u>Night</u>	<u>Date</u>
1	May 15, 1979
2	May 16, 1979
3	May 17, 1979
4	May 18, 1979
5	May 19, 1979
6	May 20, 1979
7	May 21, 1979
8	May 22, 1979
9	May 23, 1979
10	May 24, 1979

The reduced data for each block was stored on a disk called RP.PMS0451.45-XXXX, where XXXX is the name of the block. The data have been recorded also on a digital magnetic tape called PMS045.

1.3 Computer Programs for Data Reduction

In order to analyze the data and to plot the values, a set of two computer programs called FULBAR and SCAPOR are used. The basic parameters necessary to analyze the data must be given to the FULBAR program, they are:

- N = Number of data digitized (-12)
- TPD = Timing pulse period (ns)
- PIA = Number of pulse intervals averaged
- P = Pulse period (ns)
- PLOFF = Plot offset
- MAG = Plot magnification factor

The SCAPOR program, including the FULBAR program parameters, takes the first group of pulse intervals to be averaged (PIA) from the data of a particular block and calculates the overall average and the standard deviation. These two quantities, and the average for each bin, are plotted. The program then takes the second set of intervals from the data and does the same as before. Also the first set of intervals is averaged with the second set, computing a new average (combined) for each bin, an overall average, and a new standard deviation and plots the values for the bins. This process continues including each time the next group of intervals (PIA).

The first set of programs is also used only to check the quality of the data, to detect any problem and to obtain the total overall average for a block which will be used in a stopped run.

The second set of programs called FULBINO and SCAPINO is similar to the already described FULBAR and SCAPOR, but

this time, the data is taken from disk (where it has been previously stored), produces a magnified plot and punches cards with the final values for each bin, the total block average and the final standard deviation. This set of cards is used later on, to stack 2 or more blocks.

Usually after the process already described, the digitized data is recorded on the PDP-8 magnetic tape and retained from the data files.

In order to stack 2 or more blocks, the sets of cards of each block (except the first, which is used as reference) is shifted the proper number of bins so that the bins can be averaged in phase. The number of bins to be shifted on each block is given by the following relationship:

$$N_{\text{bins}} = \left\lfloor \frac{\left[T_{\text{max}} \left(\frac{T_2 - T_1}{T} \right) \right] + 1}{N} \right\rfloor \quad (2-1)$$

where N_{bins} = Number of bins the block 1 needs to be shifted with respect to block 1

T_1 = Time start sampling block 1

T_2 = Time start sampling block 2

T = Pulse period

N = Time resolution $\left(= \frac{T}{\text{Number bins}} \right)$

Frac = Fractional part

The fractional part of $\frac{T_2 - T_1}{T}$ corresponds to the fraction of the period between the time of beginning of block 1 and the time of beginning of the first block. The relationship given by 2-1 was evaluated using an HP-70C calculator.

A last program called P045 (or P10H) reads the data of the blocks to be averaged (previously shifted), calculates the average for each box weighting by the number of periods in each box, computes the final average of the combined blocks, the standard deviation σ , and plots the data. Also plotted is the 1 σ reference line, which is used as a reference in aid in deciding whether or not a pulse was present at the end of the averaging.

The A1H I program to digitize the data and the SCAPOR program were developed by Dr. John F. Slater; SCAPOR is a modification, made to adapt it to the use previously explained.

A program called DOPVEL was used to obtain the apparent pulse period for the night and time when the pulsar was observed. This program gives the pulse period at a particular date and at a given time, correcting for Doppler shift due to the earth's orbital velocity and its motion around the sun. This program was developed at the WPAO by Drs. R. B. Manchester and R. A. Ekers and made available by Dr. Steve T. Colgate.

Finally a summary of the time spent observing and reducing pulsar data and the total time of research use is presented in Table 3-1.

Table 3-1 Summary of time spent with pulsar data.

Pulsar	Optical Time Observing Pulsar	Total Time of Data Reduced	Total Time Digitizing Data
PSR1222+12	20 ^h 42 ^m	4 ^h 20 ^m	12 ^h 04 ^m
PSR2015-08	1 ^h 45 ^m	1 ^h	12 ^h

CHAPTER 3 DATA ANALYSIS

3.1 P081130-14 Data Analysis

From the 8 nights of data collected for polar P081130-14 on 3 KHz, F2-layered, pulses were found with no ambiguity on 2 nights. Pulses were found on April 18 and May 15, 1978, over averages of 100 and 150 periods respectively; the results are plotted as figures 3-1 through 3-7.

On April 18, after averaging over 100 periods ($9.0^{+1.4}^{\circ}$) a pulse appears in bin 33 with an amplitude of 1.3σ above the noise value (Figure 3-1). The 2 data blocks P112004 and P112005 used to obtain this final block have been plotted separately in figures 3-2 and 3-3. It can be seen that a pulse appears in both blocks at the same location (bin 33). Even though the pulse appearing in each block would not alone be considered to be statistically significant, the fact that they appear in the same locations for two separated sets of data provides an excellent test of the validity of the pulse.

On May 15, after averaging over 150 periods, a pronounced pulse was obtained in bin 28 with an amplitude of 3.4σ ; this pulse is shown in figure 3-4. The same test for validity as before was made. Blocks P112004 and P112005 are

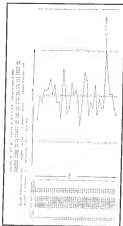


Figure 3-7 Poliovirus Hemagglutination and Poliovirus Hemagglutination Inhibition recorded as 10, 20, 30, 40, 50, 60, 70, 80, 90, 100 periods.



Figure 3-2. Average of 100 periods of signal recorded at 24.2 km on April 20, 1976. Average over 100 periods.



Figure 2-3. Pulse train recorded at 10,000 Hz on April 28, 1956. Average over 100 periods.

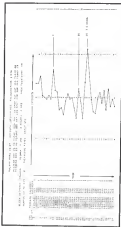


Figure 3-1. Polarizing microscope (diode excitation and HiPower) recorded at 10.5 microns on May 25, 1979. Average over 200 periods.

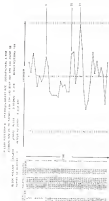


Figure 5-5
Printer Model 130-18: Message 0112540 received at 11:25:00 on May 18,
1976. Answer sent 00100000.

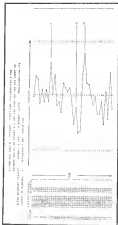


Figure 3-4. Pulse resistance index (PRI) recorded at 25, 3 days on May 25, 1978. Average over 100 periods.

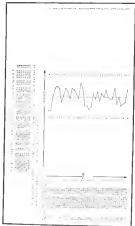


Figure 3-7. Ratio of the maximum to the minimum value of the function $f(x)$ over the interval $[0, 1]$ as a function of x .

plotted separately in figures 3-3 and 3-4. A pulse appears in bin 18 in block FIL264, with an intensity $3.6 \pm$ above the noise value, also in block FIL266, the pulse is weaker, being only $2.1 \pm$, but is at the same location (bin 18). Once again we have the fact that 2 pulses appear in the same location in two independent sets of data.

From a previous knowledge of the pulsar period and the elapsed time between the April 19 and May 15 appearances of the pulse, it should be possible to preserve its phase so that it would reappear in the same bin on the second data set it occupied on the first. However, this could not be done because of the period uncertainty. This problem will be addressed later when assigning residuals to data.

The pulse was detected in only one beam on each of the two nights, disappearing when the beam was switched to the next position. This phenomenon will be discussed later in the conclusions, when polarization effects are considered.

Before other tests made on this pulse are discussed, it is interesting to note that in bins 9 and 11 of block FIL264 (figure 3-3) two more small pulses can be distinguished. Although neither of them alone would be considered statistically significant, they both seem to appear again in block FIL266 (figure 3-4) in the same positions. The pulse in bin 11 leads the main pulse in bin 18 by 128 msec (resolution ~ 10 msec/beam) or approximately 18° of phase.

The pulse in bin 9 is separated by 488 msec from the main pulse, which is about 180° for $0.41 \pm$, $\tau = 1.40023$ sec

The pulse in bin 21 suggests the existence of a subpulse leading the main pulse. Several authors have indeed reported for this pulse a complex main pulse composed of 2 subpulses, the leading subpulse decreasing in intensity with respect to the trailing subpulse as the frequency decreases (4, 7, 8). The separation between the two subpulses has been measured at higher frequencies and extrapolation from that data suggests a much slower separation than for low frequencies than the separation obtained as 18.3 MHz between the pulses in bins 21 and 23. In 1972 Bruck and Ustinovskii (4) reported some observations made at 25 MHz with a linearly polarized antenna. They found that this pulsar shows a variety of pulse shapes with rapid transitions from one to another. In one peak, they show a pulse composed of 2 subpulses and separated by ~ 140 msec. Despite the poor resolution used (21.6 or ~ 14 msec), the 2 subpulses are clearly defined.

The separation of the subpulses as a function of frequency has been plotted in Figure 3-8, including the results by Bruck and Ustinovskii and our results (i.e., for the bin 21 and 23 subpulses). The plot suggests that either the pulse separation increases very rapidly toward lower frequencies, or the subpulse detected as 18.3 and 25 MHz is a new one which appears only at low frequencies. In Figure 3-8 the average pulse profile obtained in this thesis is compared with that obtained by Bruck and Ustinovskii. The phase of the

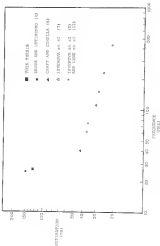


Figure 3-4: Impedance spectrum as a function of frequency (Hz)

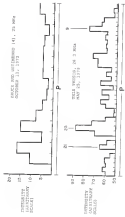


Figure 1-2. Comparison of the pulse profiles for the two observation periods obtained in the same manner of 345 periods with data obtained by direct and indirect 181 periods of 1972 period. The line of the 20.3 ms pulse profile has been shifted in order to align the peaks.

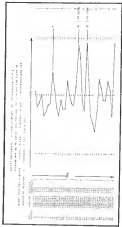


Figure 3.10. Signal waveform recorded at 10.0 kHz on May 10, 1970. Average over 100 periods.

pulse profile at 25.7 MHz has been shifted, so that the pulses in bin 23 and 25 with the pulses at 25 MHz.

Although the subpulse in bin 21 is much weaker than that in bin 23 for the 100-period average, the two are almost the same strength in the 140-period average shown in Figure 3-10 (pulses P020481). In the latter case, the heights are 1.88 σ and 1.82 σ for the bin 23 and 21 pulses, respectively. A possible implication is that the bin 21 pulse varies in intensity much more than does that of bin 23.

A small pulse appears in bin 8, when blocks P020481 and P020482 are averaged together. This pulse is also present in each block average separately (Figure 3-3 and 3-10), supporting the possible existence of an interference leading the main pulse by 0.42 μ for 150°1. In Brock and Oelkers (4) a broad and poorly defined pulse seems to appear at 0.5 μ from the main pulse, in the same plot where the double main pulse appears (Figure 3-9). So far, no interference has been reported for this pulse.

3.1.8 Peak Flux Density, Beam Flux Density and Energy for Pulsar PSR1133-04

Several continuous radio sources were observed in order to calibrate the array, in which the minimum detectable flux density, and then to obtain an estimate of the peak flux density for pulsar PSR1133-04.

The radio sources observed and their main parameters are presented in Table 3-1.

Table 3-1. Radio sources used for calibration purposes.

Radio Source	R.A.	Declination	Flux Density (Jy)	Antenna Beam Used
3C110	18 ^h 03.9 ^m	24°13'4"	382	70, 30
3C115	18 ^h 13.7 ^m	28°18'8"	133	70, 30
3C409	18 ^h 12.2 ^m	20°25'15"	181	30
3C433	18 ^h 21.6 ^m	18°34'6"	167	30

Positions and flux densities were taken from the Clark Lake Survey at 16.3 MHz (14). Of these four sources, only 3C409 and 3C433 were used in the calibrations; the 3C110 and 3C115 measurements were discarded because their difference in R.A. is only 5.8^{m} corresponding to 2.3%. Since the HPBW of the radiotelescope is 6.4° east-to-west, these two sources cannot be resolved.

The parameters of the radiotelescope for observing the calibration sources were

$$\text{Bandwidth } (B) = 100 \text{ kHz}$$

$$\text{Time constant } (\tau) = 1.5 \text{ s}$$

$$\text{Frequency } (v) = 16.3 \text{ MHz}$$

The minimum detectable flux density S_{min} , deduced from calibrations using these sources, is presented in Table 3-2.

Table 3-7. T_{min} and Galactic background temperatures deduced from calibrations

Radio Source	T_{min} K	Galactic background temp. $^{\circ}\text{K}$
30489	181	26,973
30483	87	27,840

The galactic background temperature for pulsar PSR2130+34 is

$$T_{\text{background}} = 18.51^{\circ}\text{K}$$

All the galactic background temperatures given are actual brightness temperatures, corrected for losses in the antenna.

A Hewlett Packard noise generator was used for the calibrations, and a pair of type 8720 noise diodes served as the standard of comparison for the noise generator. Details of the calibrations are given in Appendix 2.

The minimum detectable flux density has the following expression (17):

$$S_{\text{min}} = \frac{1}{\Delta\nu\tau} = T_{\text{sys}} \quad (3-1)$$

where n = number of averaged scans

$\Delta\nu$ = pre-detection bandwidth of the receiver

τ = post-detection time constant

T_{sys} = system temperature

In this case, the temperature of the system is completely dominated by the galactic background temperature, and in

equation (3-11), it is possible to calculate $S_{\text{max}}(r)$:

$$S_{\text{max}} = S_{\text{galactic background}} \approx 8$$

Assigning subscripts p to the pulsar and r to a "normal" distant radio source, it is possible by using equation (3-11), to write two similar expressions, one for the pulsar and the other for the radio source. By dividing one by the other, the following expression can be obtained:

$$S_{\text{max } p} = \frac{\frac{a_p^3 \lambda_p^2 \tau_p}{a_p^3 v_p^2 \tau_p}}{\frac{a_r^3 \lambda_r^2 \tau_r}{a_r^3 v_r^2 \tau_r}} S_{\text{max } r} K \quad (3-2)$$

The constant K takes into account that the pulsar signal, instead of being continuous, is being sampled.

Using the values obtained for the pulsar on May 18, 1979:

$$a_p = 350 \text{ periods (453 s)}$$

$$\lambda_p = 1.48 \text{ m}$$

$$\tau_p = 12.5 \text{ ns}$$

$$v_p = 18.514^{+8}$$

and the values for the radio source of

$$a_r = 1$$

$$\lambda_r = 350 \text{ m}$$

$$\tau_r = 1.5 \text{ s}$$

$$S_{\text{max } r} = 140 \text{ Jy}$$

$$v_r = 18.575^{+8}$$

the following value is obtained

$$S_{\text{max } p} = 149 \text{ Jy (considering } K = 1)$$

To obtain the peak flux density of the pulsar, it is necessary to make several assumptions

B. In order to be consistent it has been considered that the value S_{min} for both the continuous source and the pulse is that which would cause a deflection of 3σ (i.e., 3 standard deviations) from the mean value of the isotropic noise background. The 3σ criterion was used to determine whether or not a pulsed pulse was present at the end of an integration run.

B. It is assumed that the sample rate is fast enough compared with the post-detection time constant that very little information is lost in this process; then K can be considered 1.0. (Details about sampling rates can be found in Appendix B.)

C. It is assumed that the pulsed signal is unpolarized, as is the calibration source. This is probably not correct; a strong linearly polarized component has been reported (18). The effect of variable Faraday rotation in the terrestrial atmosphere would then be to cause the plane of the linearly polarized component to be oriented in different lines at any angles from parallel to perpendicular with respect to the antenna dipoles. As a result, if the polarization is actually this linear, the value of S_{min} as given by (3-2) may range anywhere from too low by a factor of two to too high by a factor of infinity. Thus, no signal at all would be detected if the linear polarization is perpendicular to the dipoles, while according to (3-2) S_{min} would not be zero.

If taking these assumptions, since the pulse reaches a value slightly higher than 1 at the end of an averaging period, the peak flux density can then be considered equal to S_{min} calculated for the corresponding number of periods averaged. Therefore,

$$S_{\text{peak } P} = S_{\text{min } P} = 158 \text{ Jy}$$

Pulse energy is usually defined as

$$E_P = \int_0^P S(t) dt \quad (3-2)$$

where P = pulse period

$S(t)$ = flux density as a function of time

For pulses with no interpulse, the limits of the integral are restricted to the duration of the pulse (pulse time duration).

In order to make the computations simpler, it is usually assumed that the pulse has a relatively simple geometrical shape. Treloars et al. (7) assumed a trapezoidal shape for most pulses (including PSR1133-14). By looking at the pulse shape obtained on May 13, 1978, one can see that this is a reasonably good approximation, considering only the pulse observed on May 15.

Using this approximation, the pulse energy in equation (3-2) can be written as:

$$E_P = \frac{1}{2} S_{\text{peak } P} S_{\text{base}} \quad (3-3)$$

where S_{base} = pulse width at base, corrected for dispersion and receiver time constant.

Substituting the values of:

$$S_{\text{peak } \nu} = 138 \text{ Jy}$$

$$R_{\text{beam}} = 71 \text{ m.}$$

the following value for S_p is obtained:

$$S_p = 3.3 \times 10^{-26} \text{ Jm}^{-2} \text{ Hz}^{-1}$$

The mean flux density of the pulse is defined as:

$$S_{\text{mean}} = S_p / P \quad (1-15)$$

Using the value obtained for S_p and the value for the period $P = 1.188 \text{ sec.}$ the following value for the mean flux density is obtained:

$$S_{\text{mean}} = 4.45 \text{ Jy}$$

It is interesting, now, to compare these values with values obtained by other authors.

In 1971, Frank and Getmantsev (14) observed the same pulsar at frequencies of 24.3, 29 and 35 MHz using a linearly polarized broadband array and reported peak flux densities in the range of 18 to 46 Jy.

In 1968, Drake and Craft (15) observed the same pulsar, using the Arecibo 1988-foot radio telescope at frequencies of 111.5 and 175 MHz and reported peak flux densities of 138 Jy at 111.5 MHz. Since these observations were made at a much higher frequency, they cannot be directly compared with the one at 24.3 MHz, nevertheless, the value obtained at the two frequencies gives an idea of the order of magnitude for the peak flux densities. A value of 38 Jy can be deduced from the data at 61 MHz given by Lovelace et al. (7).

Regarding the pulse energy and mean flux density, more recently, in 1979, Iwagawa et al. (7), have given some results at 41 MHz. They obtained a value for the pulse energy of $6.49 \times 10^{-18} \text{ Jm}^{-2}\text{sr}^{-1}$ and a mean flux density of $8.75 \times 10^{-28} \text{ Jy}$, using a linearly polarized antenna.

In 1971, Kise and Hill (3), using a circularly polarized antenna, reported at 40 MHz a value for the pulse energy of $1.4 \times 10^{-18} \text{ Jm}^{-2}\text{sr}^{-1}$ for one night (with a median over 4 nights of $1.8 \times 10^{-18} \text{ Jm}^{-2}\text{sr}^{-1}$).

It is not easy to compare the values obtained in this thesis with values obtained by other authors, because of the polarizations used and the different averaging times (but always given). Nevertheless, the value of pulse energy $E_p = 1.1 \times 10^{-18} \text{ Jm}^{-2}\text{sr}^{-1}$ obtained at 28.3 MHz, looks at least comparable with the value of $E_p = 1.5 \times 10^{-18} \text{ Jm}^{-2}\text{sr}^{-1}$ obtained at 40 MHz (which also corresponds to a peak flux density of $\sim 100 \text{ Jy}$).

According to Iwagawa et al. (7), PS0133+16 has a minimum in its average spectrum at $\nu_{\text{min}} = 15040 \text{ MHz}$. See'vin et al. (20), making simultaneous observations at several frequencies, found that the instantaneous spectrum changes from day to day. For several days it doesn't show a minimum in the spectrum but rather the intensity seems to increase towards lower frequencies.

This can be considered as part of the explanation of why the peak flux density, the mean flux density and the pulse energy obtained at 28.3 MHz are higher than the values

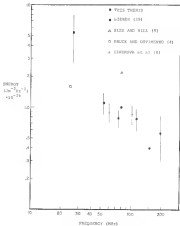


Figure 3-13. Low frequency energy spectrum for polar $\text{PbTe}(110)$.

obtained at higher frequencies; the pulses seem to show abnormally wide pulses at low frequencies derive some particular signficance, as it may have done in our case.

Another interesting aspect of this has been pointed out by Ekers and Moffat (11). They observed P80113+13 at a wave length of 13 cm (~ 1300 MHz). From a histogram of energy and number of pulses with a given energy, they noticed a long tail in the distribution, towards the high energy end. The ratio of the mean value $29.510 \cdot 10^{-16} \text{ Jm}^{-2}\text{ss}^{-1}$ to the highest energy pulse $(\sim 0.810 \cdot 10^{-16} \text{ Jm}^{-2}\text{ss}^{-1})$ is about 4.5. If one takes the value for the mean energy of $1.0 \cdot 10^{-16} \text{ Jm}^{-2}\text{ss}^{-1}$ at 24.3 MHz as an average over a relative short period (P802 16), and compares it with the calculated mean energy of $1.6 \cdot 10^{-16} \text{ Jm}^{-2}\text{ss}^{-1}$, obtained from the observations of Brock and Udaheima at 18 MHz (4) a value of 3-5 is obtained for the ratio. Such a value is not improbable according to the distribution function of Ekers and Moffat. Pulse energies obtained reported in this thesis and by several other authors are plotted as a function of frequency in Figure 3-11.

3.1.5. P80113+13 Pulse Width

Several factors can distort the actual pulse shape as observed, the factors that need to be considered are:

A. Effect of the dispersion over the finite bandwidth of the receiver

B. Effect of the post detection time constant of the receiver.

6. Polarization of the plasma and antenna

The first two factors will make the pulse wider than the actual pulse, but their effects can be corrected for if the pre-detection bandwidth, post-detection time constant of the telescope and the dispersion measure for the pulse are known, a simple form needs also to be assumed for the pulse.

In order to see the effect of dispersion let us consider the basic formula for the time of propagation through a plasma with a dispersion measure DM at a frequency ν .

The time of propagation is given by

$$\tau = \text{const } \frac{1}{\nu^2} \quad (3-6)$$

where k is a constant depending on the plasma

Differentiating this equation with respect to ν , taking finite $d\tau$ and $d\nu$ (instead of $d\tau$ and $d\nu$), and expressing the parameters in consistent units, the following expression may be obtained.

$$\Delta\tau = 4.15 \cdot 10^{-3} \text{DM} (\Delta\nu/\nu_0)^{-2} \text{ ms} \quad (3-7)$$

where DM = dispersion measure (pulses cm^{-3})

$\Delta\nu$ = bandwidth (kHz)

ν_0 = center frequency (bandwidth of DM)

This relationship gives the time it takes for a dispersed pulse to sweep through a bandwidth $\Delta\nu$ centered at frequency ν_0 .

If a triangular shape for the pulse is assumed, (3-7) is transformed into (3-4) for the half intensity level:

$$\tau_{0.5, \Delta\nu} = 4.15 \cdot 10^{-3} \text{DM} (\Delta\nu/\nu_0)^{-2} \text{ ms} \quad (3-8)$$

The correction due to the finite response of the receiver (for the half intensity level) and assuming a triangular pulse shape) is given by (8):

$$\tau_{\frac{1}{2}} = 0.443\tau \quad (13-14)$$

where τ is the post detection time constant of the receiver.

No correction can be made due to the effect of polarization loss but some discussion about its effect will be given in the conclusions.

Consequently, the corrected half intensity width of the pulse is:

$$\tau_{0.5} = \tau_{0.5 \text{ obs}} - \Delta\tau_{0.5} = \tau_{\frac{1}{2}} \quad (13-15)$$

where $\tau_{0.5 \text{ obs}}$ is the observed half intensity width.

Substituting in equations (1-6) and (1-8) the values

$$\Delta R = 4.8 \text{ ps cm}^{-1}$$

$$\Delta\epsilon = 0.8 \text{ kHz}$$

$$v_0 = 16.3 \text{ kHz}$$

$$\text{and } \tau = 11.15 \text{ ns,}$$

the following values are obtained for the corrections

$$\Delta\tau_{0.5} = 0.8 \text{ ns}$$

$$\tau_{\frac{1}{2}} = 1.8 \text{ ns}$$

Table 1-3 gives the values for the observed half intensity width and the half intensity corrected width for the pulses obtained on April 18 and May 18, 1979. The pulse width on the form is the width assuming a triangular shape, and is twice the corrected half intensity width.

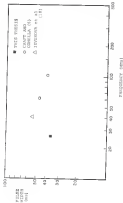


Figure 3-22 Pulse Widths and Frequency data as a function of frequency.

Table 3-3: Observed half intensity width, corrected half intensity width, and pulse width at base, for pulsar PSR1131-16

Date	$W_{0.5}$ observed (ms)	$W_{0.5}$ corrected (ms)	W_{base} (ms)
April 18, 1978	18	11	43
May 15, 1978	33	16	78

The graph of Figure 3-11 is a plot of the values of the corrected pulse half-intensity width at several frequencies (8, 81, including the values of this thesis obtained at 36.3 MHz on May 15, 1978; thus is the more reliable of the two values presented in Table 3-3.

3.4.3. Tests Performed with the Pulser Obtained on May 21, 1977

An independent and simple method for testing the pulses to see if they are indeed from a pulsar having the assumed period is to repeat the averaging process using both shorter and longer periods than that given by the B0RTH program. If the observed pulses were spurious ones which happened to have approximately (but not exactly) the same period as that of the assumed pulsar, then the averaged intensity should either increase or decrease for the shorter period, and do the opposite for the longer period. If the averaged intensity is a maximum at the assumed pulsar period, then

It is stated (approximately) that the pulsing will (continue) up to the pulse.

May 25, 1975, the date on which the best pulse was obtained, was chosen to make the test. Data were analyzed using periods longer and shorter than the assumed period by 100, 150, 180, 2100, and 2400 μ s.

Periods shorter or longer than the assumed period P by 100 μ s should cause a shift of ± 1 bin at the end of the 100 period average.

The result of this test is shown in Figures 3-13, where the relative averaged pulse intensities have been plotted as a function of the assumed period length, taking as the base the period $P = 1.18023$ μ s given by the GCPVIL program. It can be seen that the pulse intensity peaks with the assumed pulse period, and decreases sharply for shorter and longer periods. The lack of symmetry of the plot might be due to the non-symmetrical pulse shape.

The computer plots of the results for the different periods used in this test are presented in Figures 3-16 to 3-24.

The pulse intensity as a function of time has been plotted in Figure 3-25 for the pulse obtained on May 25, 1975. This pulse was chosen because it was the better of the two and is well defined. This plot is useful not only to visualize the temporal behavior of the intensity of the pulse but also to check that no interference or isolated transient pulse is mistakenly taken as a pulse pulse.

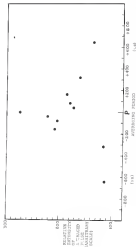


Figure 3-13 Relative Pulse Amplitude as a function of the azimuth around the polar axis (15) = 15 on May 25, 1975. $P = 1.000020$ s.



Figure 3-18. Pressure transducer data recorded on May 25, 1975, mostly and with a period 10-11 (10-11 AM).

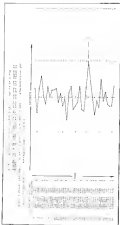


Figure 3-13. Pedestrian traffic data recorded on May 15, 1978. Recorded with a pedstat P-1 (interval = 45-100 feet).

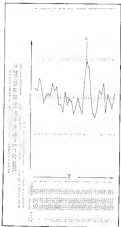


Figure 3-17. Pulse P10110-18 data recorded on May 15, 1978, sampled with a period $p = 1$ (interval $p = 100$).

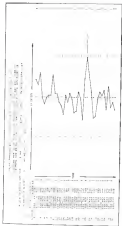


Figure 3-36. Patient P001201-18 data recorded on May 25, 1978, analyzed with a period $p = 0.387885$ s (7-18 sec).

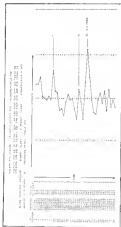
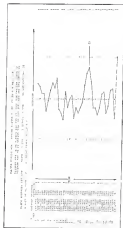


FIGURE 3-13. NOISE MAGNITUDE DATA RECORDED ON MAY 15, 1978, ANALYZED WITH A PERIOD $P = 1.000000E+000000$ (seconds period).

[illegible]

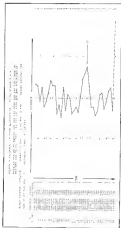


Figure 3-21. Patient temperature data recorded on May 25, 1958, multiplied with a period $P = 1.00163$ s ($P=60$ Hz).

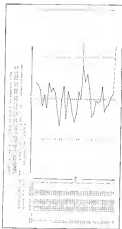


Figure 3-25: Seismic data recorded on May 25, 1978, analyzed with a period $P = 2.18812$ s (period P_0).

Figure 3-23. Pulse 19911230-16 data recorded on May 15, 1978, and (continued) on
 Figure 3-24. Pulse 19911230-16 data recorded on May 15, 1978, and (continued) on
 Figure 3-25. Pulse 19911230-16 data recorded on May 15, 1978, and (continued) on



Figure 3-23. Pulse 19911230-16 data recorded on May 15, 1978, and (continued) on
 Figure 3-24. Pulse 19911230-16 data recorded on May 15, 1978, and (continued) on



Figure 3-26. Seismic position data recorded on May 15, 1979, analyzed with a period $P = 3.149533$ s (pivot 400).

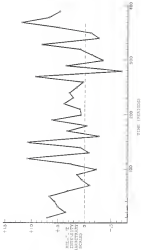


Figure 1-25 Relative pulse intensity as a function of time (periods) for pulse positions on May 14, 1979. Each point represents an average over 10 periods. $T = 1.11821$ s.

It can be recognized as a double peak (B, 198) being previously described.

Each point on the graph represents the intensity with respect to its corresponding average value of 10 consecutive periods, for his 15 (peak of the pulse at the end of the averaging process). The intensity scale is in arbitrary units and the time scale is in periods T given by the source program of $\sim 1.195831 \times 10^{-3}$ s. The plot shows that the intensity in his 15 is almost always positive with respect to its corresponding 10-periods average value, and no clear periodic variations appear over the 198 periods plotted.

3.2 PSR1943-16 Data Analysis

Of the 14 nights of data reduced for pulsar PSR1943-16, pulses were found on 4 nights: May 13, 14, 18 and 19, 1978.

A summary of the data, averaged pulse deflection in terms of standard deviations σ , and number of periods averaged is presented in Table 3-4

Table 3-4, Summary of the data, pulse deflection and number of periods averaged each night for pulsar PSR1943-16

Date	Pulse deflection in terms of σ	Periods averaged
May 13, 1978	2.04	648
May 14, 1978	2.09	616
May 18, 1978	2.28	612
May 19, 1978	2.30	612

Figure 3-19. The data for these data are presented in Figures 3-20 to 3-22. Figure 3-20 corresponds to a point at a typical day when only noise was present and no pulse appeared.

In general the pulses were of different shapes, except for the last two days, for which the shapes were similar. Some have steep leading and trailing edges (as on May 18, Figure 3-20a); others have only a steep trailing edge and a not so well defined leading edge (as on May 19, Figure 3-20b). On May 18 and 19 (Figures 3-20 and 3-20b) both pulses show a steep leading edge and much less steep trailing edge.

On May 14 the pulse was clearly present with an intensity of 3-18 σ above the mean value over an average of only 100 periods (Figure 3-20c). On the other 3 days, pulses were not seen clearly until the end of the averaging process for that night. For partial averages in these cases, some indication of a pulse can be seen at the location where the pulse appears after completing the average, but its intensity never rises significantly above the noise level. The indication of the existence of a complex pulse or subpulse can be deduced from the plotted data.

Pulsar PSR1545-14 presents at higher frequencies a complex pulse, composed of 3 subpulses, with the trailing subpulse decreasing drastically in intensity with decreasing frequency (Fig. 3-21). At 100 MHz the trailing pulse

[illegible]

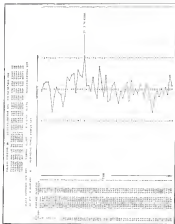


Figure 10: Total 100,000 recorded in May 10, 1970, in all four
systems over 100 periods.

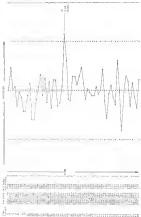


Figure 3-28. Pulse rate (b/min) recorded on day (a), (b), (c), (d) and (e).
Average over 140 periods.

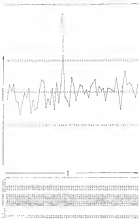
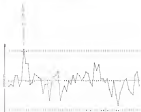


TABLE 3-10. Police 1982-83-84 recorded as 10% of 1979-80 1A plus
average over all periods.



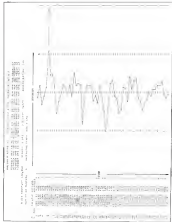
1. The signal is a damped oscillation, which is characteristic of a system with energy dissipation. The initial sharp peak represents the maximum energy input, and the subsequent oscillations show the system's response as it settles towards equilibrium.

2. The amplitude of the oscillations decreases exponentially over time, indicating that the system is losing energy at a constant rate. This is typical of a second-order system with a damping ratio between 0 and 1.

3. The period of the oscillations remains relatively constant, suggesting that the natural frequency of the system is not significantly affected by the damping. This is a characteristic feature of underdamped systems.

4. The signal eventually settles to a steady-state value, which is the equilibrium point of the system. The time taken for the signal to reach this steady state is known as the settling time, which is a key performance metric for control systems.

Figure 1-16. Palmer PDS2019-18 recorded on 8/8/19, 10/19, and 12/19
Average over 410 periods.



Volume 1-1. Survey 100142-14 recorded on May 17, 1978, at 45 mts.
Microgeometra 100 periods. No paleo present.



in the most intense, but the intensity decreases rapidly, and at 131.8 Mc is about 2/3 of the central subpulse. The intensity of the central and leading subpulse seems to remain almost constant with frequency. No indication of resolved subpulses can be obtained from the data at 45 Mc. Nevertheless, the difference in steepness of the leading and trailing edges might suggest the existence of an unresolved much less intense subpulse accompanying the main subpulse. This aspect will be discussed later when discussing the pulse width and the effect of polarization.

1.3.2 Peak Flux Density, Peak Flux Density and Energy for Pulsar PSR1504-12

In order to estimate the peak flux density of this pulsar, several radio sources were observed for calibration purposes. Radio source PSR1504-12, one of the closest to the polar, was clearly resolved despite the steepness of the galactic background (20). This radio source has apparently not been observed before at any frequency lower than 40 Mc. However, an extrapolated flux density at 45 Mc was obtained from the spectral index derived from flux density data at higher frequencies, as reported in the literature.

Spectral index can be defined (11) by the relation:

$$S = \left[\frac{1}{\nu} \right]^{\alpha} \quad (4-11)$$

where $S = 10^{-16}$ W/m² Hz.

ν = frequency

S = spectral index

For any two frequencies on the spectral part of the relationship between S and $\frac{1}{\nu}$,

$$\frac{S_{\nu_1}}{S_{\nu_2}} = \left(\frac{\nu_1}{\nu_2} \right)^S \quad (10)$$

The best available values of the flux density of the radio source 14, 15 are at frequencies of 140 and 80 MHz

$$S_{140} = 8.8 \text{ Jy}$$

$$S_{80} = 15 \text{ Jy}$$

Substituting these values into equation (9-10), a value of $S = 0.76$ is obtained.

Assuming a linear relationship for S as a function of $\frac{1}{\nu}$. Down to 40 MHz, the flux density at this frequency is found to be

$$S_{40} = 33 \text{ Jy}$$

Applying the previously used criteria, the following value for the minimum detectable flux density at 40 MHz was obtained for the region near sources PB2045-17,

$$S_{min} = 3 \text{ Jy}$$

The radio-telescope parameters for observing the radio source PB2045-17 were

$$D = 3 \text{ m}$$

$$\tau = 3 \text{ K}$$

$$n_{\nu} = 1$$

where Δv = bandwidth, τ = post-detection time constant, and n_p = number of records averaged.

It can be shown that the brightness temperature and the frequency ν are related in a similar way as λ and ν are related in equations (3-12):

$$\frac{T_{a1}}{T_{a2}} = \left(\frac{\nu_1}{\nu_2} \right)^2 \quad (3-13)$$

Using the galactic background temperatures at 45 MHz given by Ryle (24):

$$T_p = 12,800^\circ K$$

$$T_r = 7,800^\circ K$$

where T_p = galactic background temperature near the pole

T_r = galactic background temperature near the radio source PKS034-13

and the value for the galactic background spectral index given by Findlay (25), valid for the range 10 to 600 MHz:

$$\alpha = 2.45$$

the galactic background temperatures T_p and T_r at 45 MHz were found to be,

$$T_p = 7,432^\circ K$$

$$T_r = 4,432^\circ K$$

Using equation (3-3) and the appropriate parameters of the radio telescope, the minimum detectable flux density can be obtained for the polar observations.

The radio pulsar parameters were determined
(Pulsar 15-17 are)

$$\begin{aligned}S\nu_P &= 3 \text{ MHz} \\ \tau_P &= 3 \text{ s} \\ \alpha_P &= 1 \\ \tau_P &= 4.452^{\circ}\text{E} \\ \Omega_{\text{spin}} &= 8 \text{ Jy}\end{aligned}$$

The parameters when observing the pulsar are

$$\begin{aligned}S\nu_P &= 18 \text{ kHz} \\ \tau_P &= 11.25 \text{ ns} \\ \alpha_P &= \text{variable, depends on the day} \\ \tau_P &= 7.430^{\circ}\text{E}\end{aligned}$$

The peak flux density (which is also the minimum detectable flux density) for the different days is shown in Table 3-4; this table also contains the values of the mean pulse energy and mean flux density computed using equations (3-3), (3-5) and the values for the pulse width given in Table 3-5.

No known measurements of flux density and energy at low radiofrequencies exist for pulsar PSR2045-18. The lowest frequency data on this pulsar are the 80 kHz observations made by Allen and Hall (74). They attempted to observe the pulsar on 3 nights, failing to detect it and giving an upper limit for the pulse energy at $17.50 \cdot 10^{26} \text{ Jm}^{-2}\text{Hz}^{-1}$. The failure to detect this pulsar is not surprising since it has been reported to have large

Table 3-3 Summary of Peak Flow Duration, Mean Peak Discharge, Mean Flow Duration for Peak Discharge (Q_p) at 45 mm.

Date	Periods Averaged	Peak Flow Duration		Mean Peak Discharge (mm ³ s ⁻¹)	Mean Peak Discharge (mm ³ s ⁻¹)
		Q _p	Q _p		
May 18, 1979	610	1.0	1.4	1.4	1.40
May 18, 1979	610	1.0	3.4	3.4	1.40
May 18, 1979	610	1.0	3.4	3.4	1.4
May 18, 1979	610	1.0	4.4	4.4	3.25

Mean Lighted
Base

1.70

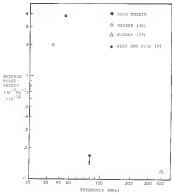


Figure 3-18 Palmer PIR2045-16 low frequency energy spectrum.

usually weak line, being often undetectable (18, 20). Our experiments at 45 MHz confirm this behavior also since it was possible to detect this pulse only on the first nights of the observing periods; it faded out for the rest of the nights.

A measured pulse energy of $41.3 \text{ Jm}^{-2}\text{sr}^{-1}$ has been given by Butler at 18 MHz (14). The lowest frequency at which published data on pulse energy is available is 418 MHz; the energy is $4.126 \times 10^{-24} \text{ Jm}^{-2}\text{sr}^{-1}$ (22, 23). Our measurement at 45 MHz seems rather high (perhaps suggesting an undetected systematic error), although no direct comparison is possible due to the lack of published information at low frequencies on this pulse.

Pulse energy as a function of the frequency has been plotted in Figure 3-32, including the data mentioned in the last paragraph.

3.3.4 Pulse Width for Pulse PSB124-18

In order to obtain the true half intensity pulse width of PSB124-18 at 45 MHz, the measured half intensity pulse width has been corrected for pulse dispersion and post detection time constant using equations (3-8) and (3-9) respectively.

In order to correct for pulse dispersion, the following values for the parameters were used:

$$F_0 = 30 \text{ MHz}$$

$$v_g = 45 \text{ MHz}$$

$$G_0 = 11.3 \text{ ps sr}^{-1}$$

The correction for pulse duration ($\Delta t_{p,5}$) using the above values is

$$\Delta t_{p,5} = 15.5 \text{ ns}$$

The correction due to the post-detection time constant of $\tau_c = 11.75 \text{ ns}$ is,

$$\tau_c = 7.5 \text{ ns}$$

The true (or corrected) half intensity pulse width $\tau_{0,5}$ is as before [equation 3-18]:

$$\tau_{0,5} = \tau_{0,5} \text{ obs} - \Delta t_{p,5} - \tau_c$$

The result for the 4 days is shown in Table 3-4

Table 3-4. Observed and corrected half intensity pulse width and pulse width at base (assuming triangular shape).

Date	$\tau_{0,5} \text{ obs}$ (ns)	$\tau_{0,5}$ (ns)	Width at base (ns)
May 12, 1979	64	31	62
May 14, 1979	58	25	58
May 18, 1979	64	31	62
May 19, 1979	64	31	62

The lowest frequencies at which pulse shape and width measurements have been given for this pulsar are 180 and 185 MHz (18, 22). Lyne et al. (12) have given the separations between subpulses at a number of frequencies, the values are plotted in Figure 3-11. The separation between

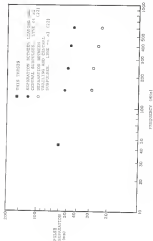


Figure 3-10 Pulse Repetition Frequency as a function of frequency, including the pulse width at the time obtained at 45 MHz as data source.

the leading and central subpulses and the trailing and central subpulses is plotted separately.

As in a previously mentioned case, the trailing pulse seems to decrease very rapidly in intensity toward lower frequencies (20), and it is likely that at 45 MHz it no longer exists or at least it can be very weak. The mean value at 45 MHz for the pulse width at its base (obtained by measuring the width at half intensity and assuming a triangular shape) of 43 ns has also been plotted in Figure 3-33.

The plotted value for the pulse width at its base, lies close to the extrapolation of a straight line passing through the points representing the separation between the leading and central subpulses. A possible conclusion of this is that the pulse observed at 45 MHz is composed of two unresolved pulses which might correspond to the leading and central subpulses at higher frequencies; the pulse width at its base would correspond to the separation between the two subpulses.

No polarization measurements have been made for this pulse at low frequency. The lowest frequency at which polarization measurements are available is 405 MHz (32, 33). The mean percentage of linear polarization at 405 MHz is 40%, and it changes during the course of the pulse, reaching about 70% for the first subpulse. The stability of the linear polarization position angle is low for the wings of the integrated pulse profile, due to the occasional

occurrence of pulses with orthogonal polarizations with respect to the mean, but as high at the center of the pulse (18). If the degree of linear polarization of the pulse is higher at lower frequencies, there will probably be a distortion in the pulse shape when it is received with a linearly polarized antenna. This distortion will depend on the variation of the polarization angle within the pulse. Assuming that the linearly polarized component in the central part of the pulse is parallel (or nearly parallel) to the antenna, then the integrated pulse profile will appear widest in the center; however the slopes of the leading and trailing edges will be weaker and the averaged pulse profile will be narrow. More details about the effect of the degree of linearly polarized component on the pulse profile and apparent intensity variations are given in the conclusions.

3.3.4. Period of Pulsar PSR2045-14 from Observations on Two Consecutive Days

The apparent mean pulse period of a pulsar can be calculated from the elapsed time between two pulses, in order to increase the accuracy it is desirable that the elapsed time between the two pulses be much longer than the pulse period. An attempt was made to obtain the apparent mean pulse period from observations made 24^h apart (elapsed time $\sim 24^h$) for pulsar PSR2045-14. On two occasions the pulsar was detected on two consecutive days,

the first encounter was May 15 and 16, 1978, and the second May 18 and 19, 1978.

The pulse period can be obtained by subtracting the epoch of appearance of the pulse x on two consecutive days and dividing by an integer number n .

The epoch of the pulse can be calculated from the relationship

$$x = T_{\text{start}} + n \cdot P \quad (3-14)$$

where:

T_{start} = time start, epoching

n = bin number where the pulse occurs

P = plotting time resolution

The information related to the two pairs of days has been summarized in Table 3-7.

The mean apparent period can be calculated from the relationship

$$P = \frac{\Delta x}{n} \quad (3-15)$$

where Δx : difference in epoch of pulse appearance on two consecutive days (elapsed time between pulse appearances)

The mean apparent periods obtained with this method are compared with the known periods in Table 3-8.

TABLE 3-7. Stock and dividend requirements when exercising the right

Date	n	$\frac{B}{(1+i)}$	i	$\frac{DB}{(1+i)}$	$\frac{B}{(1+i)}$
May 15, 1978	27	\$ 0.0128	0.4^B 4.0^B 39.5^B 39.5^B	0.4^B 4.0^B 39.5^B 39.5^B	10000
May 15, 1978	8	\$ 0.0128	0.4^B 4.0^B 39.5^B 39.5^B	0.4^B 4.0^B 39.5^B 39.5^B	10000
May 18, 1978	8	\$ 0.0128	0.4^B 3.0^B 38.5^B 38.5^B	0.4^B 3.0^B 38.5^B 38.5^B	10000
May 18, 1978	6	\$ 0.0128	0.4^B 3.0^B 38.5^B 38.5^B	0.4^B 3.0^B 38.5^B 38.5^B	10000

Table III. Comparison of the mean apparent pulse periods for PSR2218-18 obtained from observations with the OBYEL periods.

Date	Mean apparent period (s)	OBYEL period (s)	ΔP (s)
May 13, 1979			
May 14, 1979	1.84137547	1.841383	+8
May 18, 1979			
May 19, 1979	1.84137847 or 1.84140311	1.841386	+18 +14

It can be seen that the periods do not agree.

Fortunately, a jump in the epoch was noticed in the quartz time standard of Ralpin, after the first pair of days that makes the figures for the last pair of days (May 18 and May 19, 1979) very unreliable.

In any case, this discrepancy, at least for May 13 and May 14, demonstrates that the OBYEL period cannot be used to bring in phase pulses detected on two consecutive days (within 20 s limit), the apparent mean period for each day needs to be known with an accuracy of at least

$$\Delta_{\text{O}} = \left(\frac{P}{N} \right) / N \quad (13-14)$$

where N = integer number of periods between two apparitions

Expressed in terms of the elapsed time T and the period P , (13-14) becomes:

$$\Delta_{\text{O}} = \frac{P}{1 \pm T} \quad (13-15)$$

where $N = \frac{W}{P}$;

Assuming an elapsed time of 24^h (86400 s), a resolution of 0.0015 s and a period of 1.86136 s, the following accuracy for the period is obtained:

$$\Delta_{\Delta} = 4.16 \text{ ns}$$

This accuracy should bring the two pulses in phase within $2\pi/5$ bins.

It is not clear what caused the discrepancies between the CORVEL period and the period deduced from the observations; it could be an instrumental problem or a precision problem in the calculations in the CORVEL program: both possibilities were checked but the result was inconclusive. Further investigations need to be made.

3.3.4 Pulse P087951-14 Pulse Intensity as Function of Time

Figure 3-34 is a plot of the relative pulse intensity as a function of time for the pulse obtained on May 14, 1979; this pulse is the strongest and best defined. Each point plotted represents the relative intensity with respect to the noise average, averaged over 20 periods; the time axis reads in pulse periods. It can be seen that the pulse has large intensity variations: Most of the time it has a positive value with respect to the noise average, and it seems to show an almost periodic intensity modulation of ~ 100 periods (peaks at periods 50, 150 and 250).

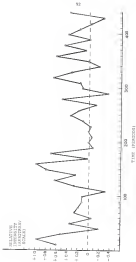


Figure 3-36. Relative pulse intensity as a function of time (periodic) for pulse frequency = 18 m/sec, 24, 180. Each point represents an average over 10 periods $\tau = 1/40000$ s.

CHAPTER 4 CONCLUSIONS

Two pulsars have been successfully detected using narrow bandwidth and long averaging times. Pulsar PSR1515-15 was detected with the University of Florida large array at 34.3 MHz and pulsar PSR2045-14 was detected using the Hedge Media Observatory large array at 88 MHz. Pulsar PSR1515-15 was strong enough to be considered statistically significant (quasi deflections larger than 3 standard deviations) in the data of 3 nights (out of 12 nights reduced). Pulses of PSR2045-14 were found on 4 nights (out of 12 reduced) with enough intensity to reach the 3 standard deviation level. Typical averaging times for both pulsars were in the range 300 to 400 periods.

Both pulsars show large stochastic variations on a time scale of days. Typically, one would appear after an averaging time as short as ~ 100 periods, and then would fade, being undetectable for averaging times as long as ~ 1500 periods over succeeding days.

The peak pulse flux density measured for PSR1515-15 is about 150 Jy, for PSR2045-14 the peak flux density is in the range of 130 to 140 Jy. These values (which have relatively large probable error, in the order of 50%), are higher than the values of the peak flux density measured by others at

quency frequencies. They are also larger than the extrapolated flux densities from published data at higher frequencies.

Also, the pulse shape and half intensity width seem to differ from extrapolated values based on data at higher frequencies.

In order to find a satisfactory explanation of the intensity variations of the two pulsars, it is necessary to consider both previously reported intensity variations and the data available on polarization at higher frequencies.

Pulsar PSR133+14 and several other pulsars have been reported to show extremely variable flux densities both from pulse to pulse and on a longer time scale, when observed at 81.3 MHz (24). These intensity variations are not believed to be caused by interplanetary scintillation, and therefore must be attributed to the source.

Pulsar PSR2045+14 has been reported as often being undetectable for several days at 408 MHz, but when it is active it has produced pulses stronger than any other of the first pulsars discovered (25).

Besides these intensity variations, the intensity of the received signal will show larger and more frequent variations when received by a linearly polarized antenna, if the pulses are strongly linearly polarized. Manchester et al. have found that most pulsars present a relatively high degree of linear polarization (26). PSR133+14 was found to have a degree of linear polarization of 37% at

483 MHz, 50% at 46 MHz and 60% at 43 MHz, which is the lowest frequency at which polarization has been measured for this pulse (15, 36). There is a clear dependence of percentage of linear polarization upon wavelength, and the tendency is for the percentage of linear polarization to increase with increasing wavelength. Pulse F061113+14 has a complex pulse, composed of two subpulses; the percentage of linear polarization is variable throughout the pulse and reaches about 70% in the middle of it at 483 MHz. For individual pulses, the position angle of the plane of linear polarization is highly stable for all but the first component, which usually shows two preferred positions perpendicular to each other. The result of this is that the degree of linear polarization of the integrated pulse is low for the first subpulse and high for the rest (21).

Pulse F082045-34 has a complex pulse composed of 3 subpulses. Manchester et al. (22) measured the degree of linear polarization to be 40% at 483 MHz, with the polarization being even larger for the first subpulse (70%). The position angle of the plane of linear polarization varies smoothly throughout the pulse; a change of about 36° occurs between the first and third subpulses. This pattern of position angle variation within a pulse is stable for many consecutive pulses. No known polarization measurements are available at lower frequencies for this pulse.

Even if the linear polarization position angle were stable when measured at a point above the terrestrial

ionosphere, the ionosphere and ionosphere relatively, large variations in the angle as measured at ground level. This is due to variations in the amount of Faraday rotation in the ionosphere. This effect is particularly strong at longer wavelengths because the angle of rotation θ is proportional to the square of the wavelength λ :

$$\theta = RM\lambda^2 \quad (4-1)$$

where RM = rotation measure of the ionospheric path.

The RM depends linearly on the electron density, N , the component of the earth's magnetic field intensity which is parallel to the direction of propagation, B_z , and the thickness of the ionosphere, h :

$$RM = \int_0^h N_e dB_z \quad (4-2)$$

It can be shown that the percent variation of θ is equal to the percent variation of the electron density N , if the thickness h and the magnetic field B_z of the ionosphere remain constant; that is:

$$\frac{\Delta\theta}{\theta} \approx \frac{\Delta N}{N} \quad (4-3)$$

It can also be shown that $\frac{\Delta\theta}{\theta}$ equals $\frac{\Delta B_z}{B_z}$ when N and h are constant, and equals $\frac{\Delta h}{h}$ when N and B_z are constant.

Kendrick has obtained values of ionospheric RM ranging from 5.1 to 6.6 rad m^{-2} , depending on the time of day and the declination of the pulsar [7]. Taking these two extreme values, it is possible to obtain the change in rotation angle, $\Delta\theta$, for let us say, a 1% change in RM as

both 26.3 and 45 MHz. The result of this calculation is shown in Table 4-1.

Table 4-1: Variations of the position angle ϕ due to variations of δ in $\delta\theta$, $\delta\theta_0$ or δL in the ionosphere.

Frequency (MHz)	ϕ for $\frac{\delta\theta}{\theta}$, $\frac{\delta\theta_0}{\theta_0}$ or $\frac{\delta L}{L} = 1\%$, and $\delta\theta = 0.1$ rad $m^{-1/2}$	ϕ for $\frac{\delta\theta}{\theta}$, $\frac{\delta\theta_0}{\theta_0}$ or $\frac{\delta L}{L} = 1\%$, and $\delta\theta = 1$ rad $m^{-1/2}$
26.3	7.4°	44°
45	1.2°	180°

From the values obtained in Table 4-1, one should expect large random (or nearly random) variations of the position angle for large values of $\delta\theta$ at the two frequencies considered. This effect could produce a modulation which distorts the pulse shape, in addition to the long term intensity variations. For low and medium values of $\delta\theta$, the variations of the rotation angle ϕ are not as important, since they are small. There is another effect that can be important for large values of $\delta\theta$: that is the change in position angle when observing the pulsar in different regions of the sky (i.e., when observing the pulsar with different beams), or when observing the pulsar on different nights. For example, at 45 MHz, with a $\delta\theta = 1$ rad $m^{-1/2}$, a change of $\delta\theta$ of π , $\pi/2$ or $\pi/4$ (or a combination of these), will cause the position angle to be orthogonal to its previous position. If the pulsar

Radiation is relatively highly linearly polarized, and should produce a significant drop in the intensity of the received signal when using a linearly polarized antenna. This last effect can provide a possible explanation of the disappearance of the polar signal when switching to a different beam of the 18.3 MHz array.

Regarding the effect of the polarization throughout the pulse, the pulse shape can be modified when using a linearly polarized antenna; the signal will appear more intense for the part of the pulse with polarization that matches the polarization of the antenna, and abnormally weak for the part of the pulse with positions angle unfavorable for the antenna. Also if a given part of the pulse presents random changes in the position angle from normal to orthogonal which led to a low integrated polarization for that part of the pulse, the intensity of the averaged pulse will be abnormally weak in that part. This seems to be the case with F8B113456, in which the observed intensity at 24.3 MHz of the second subpulse is usually stronger than the intensity of the first subpulse, the latter is often undetectable or appears only occasionally. The separation of these two subpulses at 24.3 MHz is much larger (700 ns) than the predicted separation for the two subpulses which compose the main pulse) obtained from data at higher frequencies. However the separation at 24.3 MHz does agree with the results obtained at 18 MHz (40).

The values for the peak flux density, the mean flux density and the mean energy are higher than the values obtained from extrapolation at higher frequencies, with the exception of a value of peak flux density of 180 Jy obtained at the relatively high frequency of 111.5 MHz for FR1113-14. Lovelace et al. (7) and Kar'yan et al. (20) made simultaneous observations of this pulsar at several frequencies in order to obtain the spectrum. They found that the average spectrum (average over several nights) presents a turn-over at about 40 MHz. Nevertheless they also found that the spectrum shows large variations from day to day in the range 40 - 100 MHz, often with a change in the sign of the spectral index. It is interesting to note that the spectrum of some individual nights does not show a turnover but rather an increasing intensity with decreasing frequency. This might explain (at least in part) the unusually large intensity of the pulsar at 26.5 MHz, when compared with average values obtained by other authors at higher frequencies.

Averaging of pulsar for two different nights was not possible. For pulsar FR13045-14, the period obtained from the GUPPI program differs from the period deduced from observations on two consecutive nights by amounts between 8 to 28 μ s. This discrepancy is large compared with the required precision for averaging over two consecutive days (i.e., 4 weeks of μ s). No attempt was made for FR1113-14 to average data of two different nights since the only two

nights with data are 17 days apart. It is not clear up to the present whether the discrepancy of the periods is due to instrumental problems or insufficient precision in the calculations of the predicted period.

From the results obtained in this thesis, it has been shown that it is possible to detect pulsars at both 16.3 MHz and 45 MHz, with certain limitations. In the light of the results obtained, some improvements and modifications in the equipment and data reduction can be suggested which will make it possible to obtain even more information about the received pulsars.

One such modification is already being put into effect, a greatly improved unmodulated 16.3 MHz receiver which will replace the old receiver is under construction and is nearly completed.

Recording at a much wider bandwidth (perhaps 30-100 MHz), and utilizing a de-dispersion technique, would considerably increase the signal-to-noise ratio. The wide band signal would instantly be recorded using an instrumental tape recorder of suitable bandwidth. It could then be played back into the Bruel Kjaer Fine Spectrum Analyser, the output of which could be digitized and further analysed by means of a microprocessor or a minicomputer (like the PDP 1134). After the data has been converted to a power spectrum appearance in the 500 output frequency channels of the Bruel Kjaer spectrum analyser, the frequency dispersion can then be removed by means of the computer. With a 100 MHz-bandwidth

system of this type, the signal-to-noise ratio would be increased by a factor of $\sqrt{\frac{100}{1}}$, or 10, over what it was with the 8 kHz-bandwidth system.

Another advantage of a wide bandwidth system, and perhaps an even more important one, is its ability to determine the degree of linear polarization of the pulsar signals by taking use of Faraday rotation in the ionospheric ionosphere. As stated previously, the rotation of the plane of linear polarization varies, due both to ionospheric fluctuations and to a continual change in the ray path as the earth rotates, so that the polarization plane is frequently parallel and frequently perpendicular to the antenna dipoles. The degree of linear polarization is readily calculable from the measured intensities at times of parallel and perpendicular orientations. This appears to be an excellent method for determining linear polarization. It has apparently not previously been used for pulsars.

Also, if the rotation measure (RM) of the ionosphere above the observatory can be determined independently, the position angle of the plane of linear polarization of the pulsar signal at the top of the ionosphere can be calculated.

It may also be possible to obtain still more information about polarization. Since the dipoles of the 16.3 MHz array at the Davis County radio observatory have the capability to be rotated to any position, half of the array can be rotated in 90° respect to the remaining half, making possible to obtain information about the circularity

polarized component of the emission. If the cone (plane) there is enough signal-to-noise ratio, it would also be possible to set the antenna in such a way that one third of the array is oriented at 0° , one third at 45° and the last third at 90° , from the information obtained in this configuration it is possible to deduce the Stokes parameters of the incident wave. Any polarizer polarization measurements made at 28.3 MHz and 49 MHz would be useful since none have been made at frequencies lower than 49 MHz (34).

Improvements need to be made in the calibration in order to obtain more accurate values for the peak flux density. This can be accomplished by simulating the polar pulse using a precise time standard and a frequency synthesizer, in order to obtain periods within some 1 second of the observed period. It should be possible to obtain estimations of peak flux densities by digitizing and processing this signal in a similar manner to the actual pulse. It will be also necessary to find out the reasons for the discrepancy between the GUPVL period and the period obtained from the observations, simulations with an accurate period should help in determining if the discrepancy is due to instrumental problems. A more careful analysis of the accuracy in the calculations of the GUPVL program needs to be made. The solution of this problem will make it possible to average several nights and obtain the long-term average pulse characteristics.

APPENDIX A
LIST OF PULSARS

Tables A-1 and A-2 contain the pulsar candidates to be observed using the University of Florida 34.3 MHz array and the 45 MHz array at the Ridge Radio Observatory.

Table A-3 gives the references from which the data at the different frequencies were obtained.

Table A-1. Polar coordinates of the observed water/ice interfaces for 10 comets
20-1 Nov. 1976.

Polar	Period	$(R_1 - R_2)$	ΔT_0	Polar Energy $\sim 10^{-11}$ $2\pi \times 10^{-11}$	Mass Density $\sim 10^{-28}$ g	Polar width half degeneracy (m)	Observed half- ground range (m)	Length Angle at transit ($^\circ$)
0021-07	0.943	20	24.8	8.00 # 61.000	1.2 # 61.000	12 # 100.000	12,000	45
0028+74	1.293	8	33.9	3.000 # 61.000	3.3 # 61.000	125 # 61.000	64,000	44
0034+06	1.074	13	48.1	3.000 # 61.000	3.1 # 61.000	34 # 61.000	34,000	34
1120+16	1.100	3	14.05	1.000 # 61.000	0.75 # 61.000	47 # 61.000	27,000	14
1608+00	0.730	20	73	1.000 # 61.000	1.00 # 61.000	10 # 100.000	20,000	26
1908+01	1.327	13	45.3	3.000 # 61.000	3.8 # 61.000	20 # 100.000	38,000	9

Table A-1. List of pulsar candidates to be observed with Radio (radio observations)

Pulsar	Period (s)	DM $\pm \Delta$ (pc cm ⁻³)	Δt_D # 10 km (ms)	Pulsar Energy 10 ³¹ W $\pm \Delta$	Plan Intensity 10 ⁻²⁶ W $\pm \Delta$	Pulsar Height 10 ¹² m Determined (km)	Galactic North Galactic Plane (°)	Latitude Azimuth as constant (°)
0011-07	0.00209	10.00	5.5	500 # 50000	1.5 # 5.000	50 # 300000	~0,500	0
0020-20	1.2443	34.30	16.3	2,100 # 10000	1.5 # 31.000	105 # 81.000	~0,400	0
0020-33	1.02046	53	47.4	300 # 400000	0.5 # 10	270 # 400000	~0,400	0,8
0031-06	1.07754	15.00	11.1	3,100 # 51.000	0.5 # 10	30 # 51.000	~0,500	30
1515-20	0.30200	60.00	46.3	150 # 60000	1.0 # 34.000	~30 # 500000	~30,000	0
2015-18	2.30057	15.34	10.47	0.5 # 10	~1 # 34.000	~0.5 # 50000	~0,500	0

(*) Information not available at that frequency.

Table A-1 List of references from which the data at the different frequencies were obtained.

Frequency MHz	References
14	Rucker (30)
18	Rucker (26)
41	Iveskov et al. (8)
80	Blas and Hall (9)
100	Iveskov et al. (7)
151	Lyne et al. (22)
400	Rosenblatt et al. (18) and Renshaw et al. (20)

APPENDIX B SAMPLED RATE, TIMING PULSE FREQUENCY AND DATA POINTS PER SECOND

For the configuration of the data reduction system used (Figure 1-15), the sampled values are sent to the Ansul 450/WS computer in real time. In that condition the rate of transmission of data into the Ansul is the limiting factor for the sampling rate.

The absolute maximum transmission rate to the Ansul 450/WS is 1000 baud (20 characters/second). The maximum useful rate is 80 characters/second (40 characters/second as a more conservative but reliable rate). Since the A/D converter has 12 bits, which average 3 characters per data point, the transmission rate will be $40/3 = 13$ data points per second.

The number of data points per second (DPS) obtained from the RIM I is given by the relationship (B-1):

$$DPS = \frac{1}{TTP \cdot NVALS \cdot TRAMP} \quad (B-1)$$

where TTP = RIM I timing pulse period (ms)

NVALS = Number of sampled values averaged before sending to the terminal

TRAMP = Parameter for the RIM I program

The TTP for the RIM I must not be shorter than 4 ms (250 Hz); the value used in the digitizing process for the

pulses was 4 no. [125 Hz]. The number of serviced pulses averaged (SAMP) before sending through the channels to the Analog is fixed at 8.

Table B-1 gives the relationship between the TPP, TRAP and the DPS (18).

Table B-1. Values of DPS as a function of TPP and TRAP.

TPP (Hz)	TRAP (Hz)	DPS				
		01	02	03	04	05
4	350	11.85	15.415	20.4102	7.8125	4.250
8	125	15.015	2.8015	4.2003	2.800	2.125

Data Points Per Second (DPS)

In order to meet the transmission rate requirements of 12 data points per second using 4 no. for the TPP, a safe value of TRAP = 02 must be used, that gives a value of 7.8125 DPS.

If the tapes are played back at the same speed that they are recorded, then the resolution would be also 7.8125 DPS, which is a poor resolution. However, if the tapes are played back with a slowdown of 4:1 the time resolution will be increased by a factor of 4.

Table B-2 gives the number of DPS obtained in real time and an actual or original pulse rate (18).

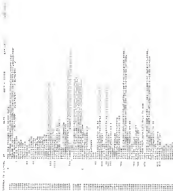
Table B-2. Sample 20 data points per second as read into original pulse time.

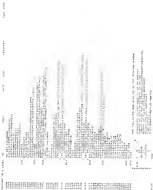
Samples per second					
Rate (cps)	10/90	10,100	7,8125	6,100	5,000
Original pulse time	1.1111	40.000	31.25	20.00	20.0000

SINCE 90% is fixed at 9, that means that 9 sampled values are averaged together before producing one data point, in that way, the tape is being sampled 9 times faster than the data point rate. As an example, if a TRF = 8 ms and a TRSR = 62 are being used, 7,8125 SPS are obtained, but the tape is being sampled at the rate of 62 S samples/second as read time, which in original pulse time is 190 samples/second.

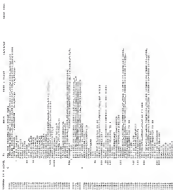
APPENDIX 1 ACRONYMS

INDEX 3 COMPUTER PROGRAMS





ACAPHS Program



2015-10-15 10:10:10

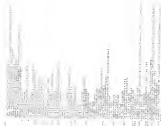


Figure 1. Distribution of chemical elements across different categories.



Fig. 1. Relationship between the logarithm of the number of species and the logarithm of the number of individuals.

APPENDIX B 34.3 MHz ARRAY CALIBRATION

In order to obtain the minimum detectable flux density and then estimate the peak flux density for pulsar PSR1509-18, several calibrations were performed with the array.

The standard noise source was provided by a Synthesizer 8712 noise diode; as a secondary noise source a Hewlett Packard amplifier was used. Several radio sources of known flux density and the galactic background temperatures were also used in the calibrations.

Figure B-1 is a block diagram of the 34.3 MHz array signal transmission and calibration system. Usually the calibration signals are introduced at point ① (Figure B-1). In order to improve the calibrations, the signal was later introduced at point ② (Figure B-1); by doing this, any uncertainties in the gain for losses of the 18 preamplifiers and the 2 S-B Butler matrices are avoided.

Two 1 input-by-8 outputs 50-ohm power splitters were used to split the calibration signal into the 18 preamplifier inputs. To properly match all the outputs, two power splitters were loaded with 50-ohm resistors.

As a result of introducing the calibration signal at ① and then at ②, a net gain of ± 1 dB was determined for the 18 preamplifiers plus the 2 S-B Butler matrices, several

other gains and losses were also determined for the preamplifiers, 150 Mc filters and aluminum coaxial lines; all the losses and gains determined are shown in Figure 8-1.

8.1. Losses Between the Dipole and Point ③

The procedure to obtain the losses between the dipole and point ③, which includes the 10 d-w helical sections, the hybrid rings, the matching networks and the parallel transmission lines, is not as straightforward as it is to obtain gain and losses for the rest of the system. An indirect method was therefore used in this thesis. First, the receiver deflection due to the galactic background was measured. Then a signal from the noise generator was introduced at point ③ (using the indicated switch), and was adjusted to give the same recorded deflection as that from the galactic noise. From the ratio of the noise temperature of the introduced signal to that of the part of the sky from which the galactic signal was obtained, which is available in the literature, the losses can be obtained. A summary of the results is shown in Table 8-1.

The noise temperature for the noise diodes has been calculated using the following relationship [17]:

$$T = T_s + \frac{e^2 I_s}{4k} \quad (8-1)$$

where T_s = ambient temperature ($\approx 290^\circ K$)

e = electron charge ($= 1.6 \times 10^{-19}$ C)

I_s = diode's plate direct current

R = plate load resistance

k = Planck constant ($= 1.38 \times 10^{-23}$ J $^\circ K$)

Substituting in equation (E-1) the values for the constants, using the value $R = 30$ cm and expressing I in milliamperes, the following relationship can be deduced

$$T = 340 + 249.45 I \quad (E-2)$$

Equation E-2 gives the temperature in $^{\circ}\text{K}$ and it is valid for $R = 30$ cm.

Galactic background temperatures for the first two points in the sky have been obtained from Turtle at 21.440 and 24.3 MHz. The value for the third point was obtained from measurements made by Righi (23) at 34 MHz and corrected to 24.3 MHz using the galactic background spectral index given by Field (22) of $\alpha = 2.43$, which is valid for the range 10-100 MHz.

It is interesting to compare the losses obtained in Table E-1 with the losses obtained by summing the estimated dB attenuations along the path between the dipole and point (3). The sum of estimated or measured losses of the individual components is given in Table E-2 (4).

The average of the measured losses given in Table E-2 is 7.2 dB which agrees well with the losses computed from Table E-1.

Table B-2. Losses between dipoles and point ② considering the partial losses.

	Maximum dB	Maximum dB
Full wave Bell Line	.38	.38
Hybrid ring	.45	.45
Phase switch	.80	1.10
Coaxial switch line	1.80	1.70
B-W Butler matrix	1.48	1.48
Line connecting B-W to B-S Butler matrices	.5	1.00
Total	5.38	6.13

The corrected noise temperatures at point ②, given in Table B-1, is the noise temperature of the noise source corrected for the 5 dB losses in the aluminum coaxial line and the losses in the power splitter (factor 0.76); losses in the power splitter consider the losses due to the two 50-ohm resistors and the losses at 24.2 GHz of the power splitter itself.

B.2. Antenna Effective Area

By using the losses between the dipoles and point ② already obtained and calibrations done with noise source 3C412, the antenna effective area can be calculated; the main advantage of obtaining the change in antenna temperature

ΔT_a for a radio source of known flux density S). The relationship that gives the effective area is

$$A_{\text{eff}} = \frac{2k \Delta T_a}{S} \quad (3-3)$$

ΔT_a is the change in antenna temperature referred to the dipoles, corrected for losses ϵ between point (3) and the dipoles. If ΔT_a^0 is the change in temperature measured at point (3), then the actual change in antenna temperature is

$$\Delta T_a = \frac{\Delta T_a^0}{\epsilon} \quad (3-4)$$

where ϵ = transmission line efficiency, which in this case considers the losses between point (3) and the dipoles.

From the calibrations, the following values were obtained for 3043:

$$\begin{aligned} S &= 367 \text{ Jy} \\ \Delta T_a^0 &= 153^\circ\text{K} \\ T_a &= 15,808^\circ\text{K} \\ \epsilon &= 0.15 \end{aligned}$$

Using equations 3-3 and 3-4, the following value is obtained:

$$A_{\text{eff}} = 14,858 \text{ m}^2$$

This value needs to be corrected by a factor $p_1 = 0.49$, given in the Array program (11), which considers the fact that the source is off the center of the beam. The corrected effective area of the antenna is then:

$$A_{\text{eff}} = 7,340 \text{ m}^2$$

The value of the effective area can be also expressed in terms of the change in surface temperature in $^{\circ}\text{K}$, divided by the flux density as $\partial T/\partial y$, this quantity is usually denoted as λ_0^* (13),

$$\lambda_0^* = \frac{\Delta T_{\text{eff}} + p_1}{\partial T/\partial y} \quad (13-11)$$

Substituting in (13-11) the values

$$\Delta T_{\text{eff}} = \Delta T_{\text{eff}}^{\text{max}}/2 = 1553^{\circ}\text{K}$$

$$p_1 = 243 \text{ J/g}$$

$$p_1 = 5.08$$

The following value for λ_0^* is obtained

$$\lambda_0^* = 4.3 \cdot 10^3 \text{ J/g}$$

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BIOGRAPHICAL SKETCH

Emmanuel Reyes was born in Santiago, Chile, on December 8, 1940. After finishing high school and with a strong desire to study electrical engineering (in which he was encouraged by his parents, sisters and brother), he entered the School of Engineering of the University of Chile. It was there where he became interested in astronomy while taking a course with Professor Claudio Aguirre; his interest grew when he was allowed to use the old 8-inch refractor at Cerro Chile. He started his amateur astronomical activities by building his own small Newtonian reflector and spending hours and hours watching the interesting southern sky (in which he was enthusiastically supported by his two sisters Lucía and Conchita and his brother Juan).

His professional interest turned to radio astronomy when in 1974 Jorge Rey asked him to work on the 85 MHz array in Magall as a thesis project.

In 1978 he married María Angélica, and they now have two daughters: Carolina and Daniela (who just joined the family).

In 1977 he obtained a degree in electrical engineering from the University of Chile, and in September of the same year

he came to the University of Florida to continue his studies in astronomy. While studying at the University of Florida he has been teaching elementary physics laboratories, elementary astronomy laboratories and digital electronics laboratories.

His other interests are mountain climbing, Russian music and photography.

I earnestly certify that I have read this study and that my opinion is conforming to acceptable standards of scholarly presentation and is fully adequate, in scope and quality, as a thesis for the degree of Master of Science.



Thomas D. Carr, Chairman
Professor of Astronomy
and Physics

I certify that I have read this study and that as my opinion is conforming to acceptable standards of scholarly presentation and is fully adequate, in scope and quality, as a thesis for the degree of Master of Science.



Alex C. Smith
Professor of Astronomy
and Physics

I certify that I have read this study and that as my opinion is conforming to acceptable standards of scholarly presentation and is fully adequate, in scope and quality, as a thesis for the degree of Master of Science.



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This thesis was submitted to the Graduate Faculty of the Department of Astronomy in the College of Liberal Arts and Sciences and to the Graduate Council, and was accepted as partial fulfillment of the requirements for the degree of Master of Science.

December, 1966


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